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Electronics – 96032



White Noise Filtering: LPF

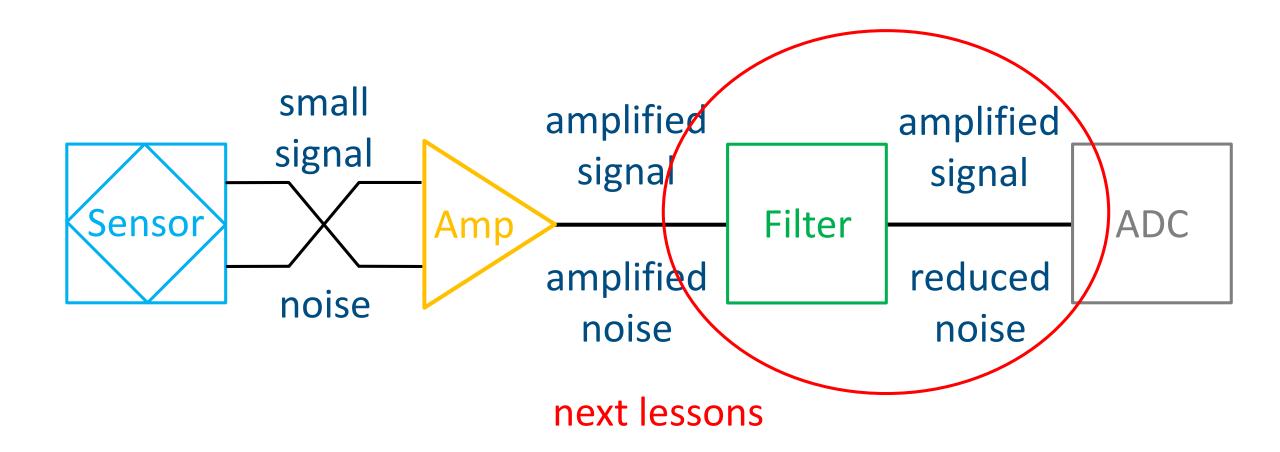
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Slides are supplementary material and are NOT a replacement for textbooks and/or lecture notes

Acquisition chain



Purpose of the lesson

- It is now time to begin a discussion on the techniques for improving *S*/*N*
- Noise-reduction techniques obviously depend on the type of signal and of noise:

| | | Noise | |
|--------|---------------|-------------|--------------|
| | | HF (white) | LF (flicker) |
| Signal | LF (constant) | this lesson | next lessons |
| | HF (pulse) | next lesson | next lessons |



- Intro WN filtering
- LPF output signal and noise in the time domain
- LPF output signal and noise in the frequency domain



- Usually placed after the sensor and the first amplification stage
- Its fundamental function is to increase the signal-to-noise ratio
- Two basic approaches
 - Time domain
 - Frequency domain

White noise filtering: concept

Consider a constant signal plus white noise

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- We know that the average value of noise is zero ⇒ to reduce noise, we can compute the time average (ergodic process)
- Note that averaging does not affect the (constant) signal \Rightarrow we improve S/N

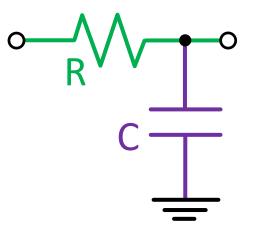


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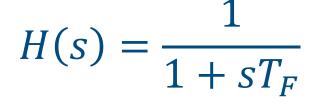


• Delta-function response

$$h(t) = \frac{1}{T_F} e^{-\frac{t}{T_F}} \mathbf{u}(t)$$



• Transfer function



• What kind of average are we performing here?

9

Signal (time domain)

Step response, x(t) = Au(t)

$$y(t) = x(t) * h(t) = \int_{-\infty}^{t} x(\tau)h(t-\tau)d\tau = \int_{0}^{t} A \frac{1}{T_{F}} e^{-\frac{t-\tau}{T_{F}}} d\tau$$

= $A \left(1 - e^{-\frac{t}{T_{F}}}\right)$ weighting function \Rightarrow
 $x(\tau)$ exponential average
 $h(t-\tau)$

Weighting function time correlation

$$k_{hh}(\tau) = \int h(t)h(t+\tau)dt =$$

$$\frac{1}{T_F^2} \int e^{-\frac{t}{T_F}} u(t)e^{-\frac{t+\tau}{T_F}} u(t+\tau)dt =$$

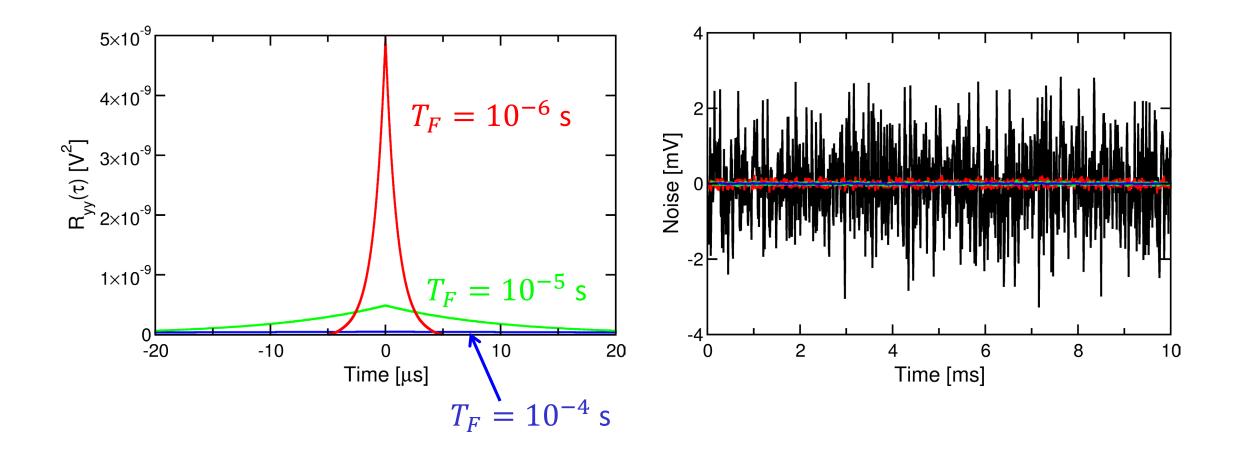
$$\frac{e^{-\frac{\tau}{T_F}}}{T_F^2} \int_0^\infty e^{-\frac{2t}{T_F}} dt = \frac{e^{-\frac{\tau}{T_F}}}{2T_F} \Rightarrow \frac{e^{-\frac{|\tau|}{T_F}}}{2T_F}$$

Noise (time domain)

$$R_{yy}(\tau) = R_{xx}(\tau) * k_{hh}(\tau) = \lambda k_{hh}(\tau) = \frac{\lambda}{2T_F} e^{-\frac{|\tau|}{T_F}}$$
$$\overline{n_y^2} = \lambda k_{hh}(0) = \frac{\lambda}{2T_F}$$

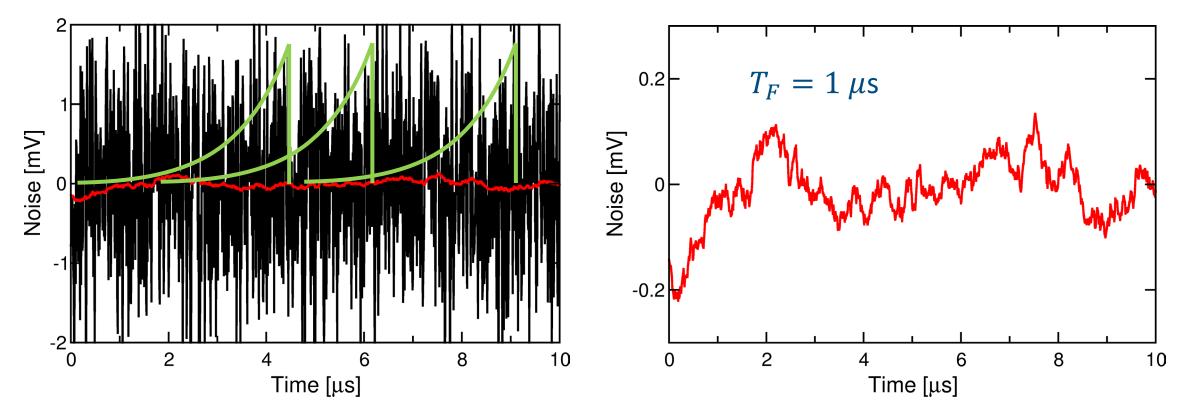
Noise power is reduced when T_F is increased (noise is averaged over a longer interval)

Noise reduction (time domain)



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Non-white (correlated) output noise



Output noise is an average of the input WN: its values are correlated (i.e., «similar») over the extent of the WF

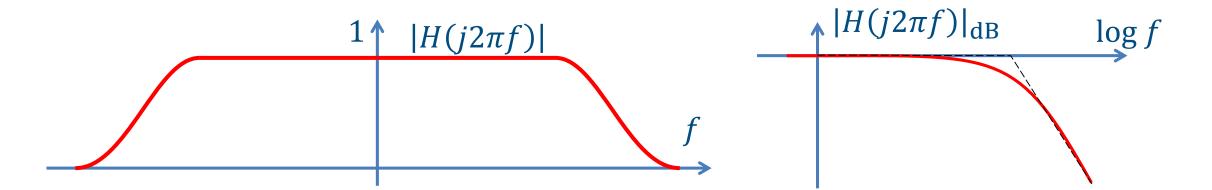


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Signal (frequency domain)

Step response:

$$Y(s) = H(s)X(s) = \frac{1}{1+sT_F}\frac{A}{s} = A\left(\frac{1}{s} - \frac{T_F}{1+sT_F}\right)$$
$$\Rightarrow y(t) = A\left(1 - e^{-\frac{t}{T_F}}\right)u(t)$$



16

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Noise (frequency domain)

$$S_{y}(\omega) = S_{x}(\omega)|H(\omega)|^{2} = \lambda|H(\omega)|^{2} = \frac{\lambda}{1+(\omega T_{F})^{2}}$$

$$\overline{n_{y}^{2}} = \int_{-\infty}^{\infty} S_{y}(\omega)\frac{d\omega}{2\pi} = \frac{\lambda}{2\pi}\int_{-\infty}^{\infty}\frac{d\omega}{1+(\omega T_{F})^{2}} = \frac{\lambda}{2\pi T_{F}}[\arctan(\omega T_{F})]_{-\infty}^{+\infty} = \frac{\lambda}{2T_{F}}$$

$$I = \int_{-\infty}^{|H(j2\pi f)|^{2}} f = \int_{-\frac{1}{4T_{F}}}^{|H(j2\pi f)|^{2}} f = \int_{-\frac{1}{4T_{F}}}^{\infty} f = \int_{-\frac{1}{4T_{F}}^{\infty} f = \int_{-\frac{1}{4T_{F}}^{\infty} f = \int_{-\frac{1}{4T_{F}}^{\infty} f = \int_{-\frac{1}{4T_{F}}^{\infty}$$

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Signal and noise bandwidths

A single-pole LP filter has:

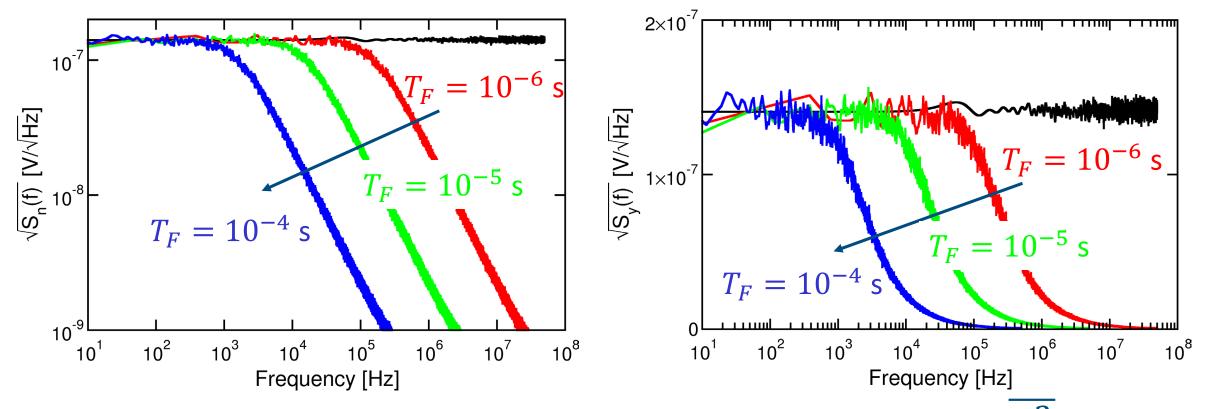
a signal bandwidth
$$BW_s = f_p = \frac{1}{2\pi T_F}$$

a (white) noise bandwidth $BW_n = \frac{1}{4T_F} = \frac{\pi}{2}BW_s$
 $1\uparrow^{|H(j2\pi f)|}$
 $I\uparrow^{|H(j2\pi f)|}$
 $BW_s BW_n$

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Noise reduction (frequency domain)



- LPF rejects high-frequency noise components, reducing n_v^2
- Noise power is reduced when BW_n is lowered

Input and output *S*/*N*

• We consider a quasi-white noise at the input, with equivalent bandwidth $f_n = 1/2T_n$

$$\left(\frac{S}{N}\right)_{x} = \frac{V_{i}}{\sqrt{\overline{n_{x}^{2}}}} = \frac{V_{i}}{\sqrt{2\lambda f_{n}}} = V_{i}\sqrt{\frac{T_{n}}{\lambda}}$$

• At the output we have

$$\left(\frac{S}{N}\right)_{y} = \frac{V_{i}}{\sqrt{n_{y}^{2}}} = \frac{V_{i}}{\sqrt{2\lambda BW_{n}}} = V_{i}\sqrt{\frac{2T_{F}}{\lambda}}$$

Improvement of *S*/*N*

